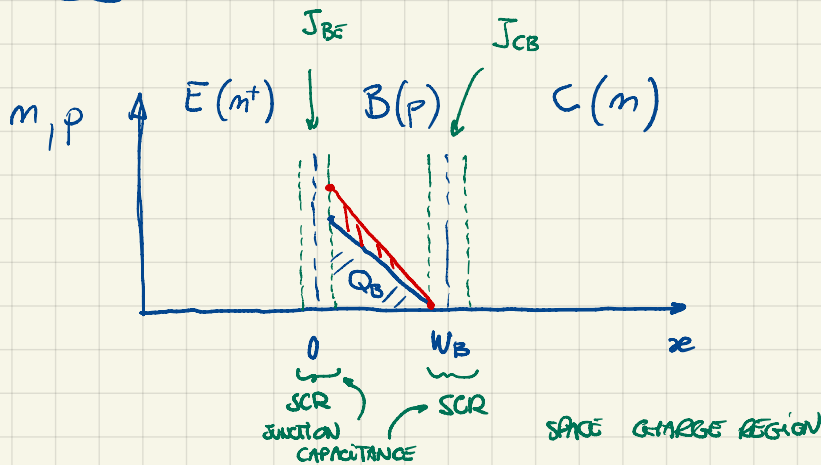


11/10/2022

CHARGE STORAGE EFFECTS IN BJTs AND MOSFETS

- CHARGE STORAGE EFFECTS ARE INTRINSIC TO BOTH BJTs AND MOSFETS
- THESE EFFECTS ARE THE ORIGIN OF ω_H

◇ BJT



IN THE FORWARD ACTIVE REGION:

J_{BE} IS DIRECTLY BIASED

J_{CB} IS INVERSELY BIASED

WE HAVE MINORITY CARRIER ACCUMULATION IN THE BASE REGION

THE CHARGE STORED IN THE BASE IS APPROXIMATELY GIVEN BY

$$Q_B = I_C \cdot \tau_F \quad \tau_F \in [1 - 100 \text{ ps}]$$

τ_F : AVERAGE LIFETIME OF A MINORITY CHARGE (e^- IN THIS CASE) INSIDE THE BASE REGION \rightarrow TECHNOLOGICAL CONSTANT

$$I_C \approx I_S \cdot \exp\left(\frac{V_{BE}}{V_T}\right)$$

WE CAN LINEARIZE THIS FUNCTION $Q_B = f(V_{BE})$ AROUND THIS DEVICE OP.

$$C_{diff} \triangleq \left. \frac{\partial Q_B(V_{BE})}{\partial V_{BE}} \right|_{V_{BE}=V_{BE}} = \tau_F \cdot \frac{I_C}{V_T} = \tau_F \cdot g_m$$

BIAS POINT CURRENT

C_{diff} IS THE MAIN COMPONENT OF THE CAPACITANCE (SMALL SIGNAL) THAT WE CAN MEASURE AT THE B-E PORT, THAT IS CALLED

$$C_{TC} = C_{diff} + C_{J_{BE}} \approx C_{diff}$$

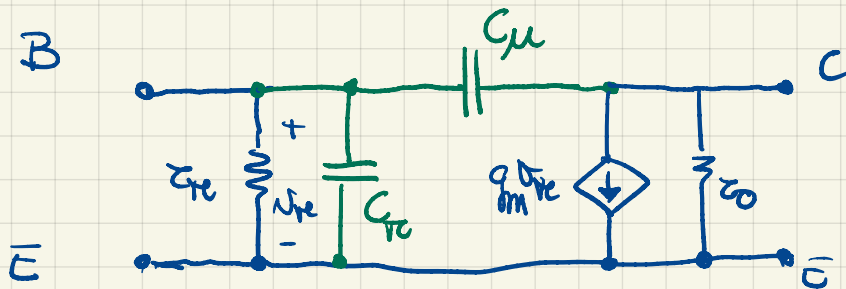
NEGLECTIBLE DUE TO J_{BE} BEING FORWARD BIASED

WE ALSO HAVE A SECOND CHARGE STORAGE MECHANISM, DUE TO SPACE CHARGE REGIONS. THIS EFFECT IS MORE RELEVANT FOR THE CB JUNCTION THAT IS REVERSE-BIASED.

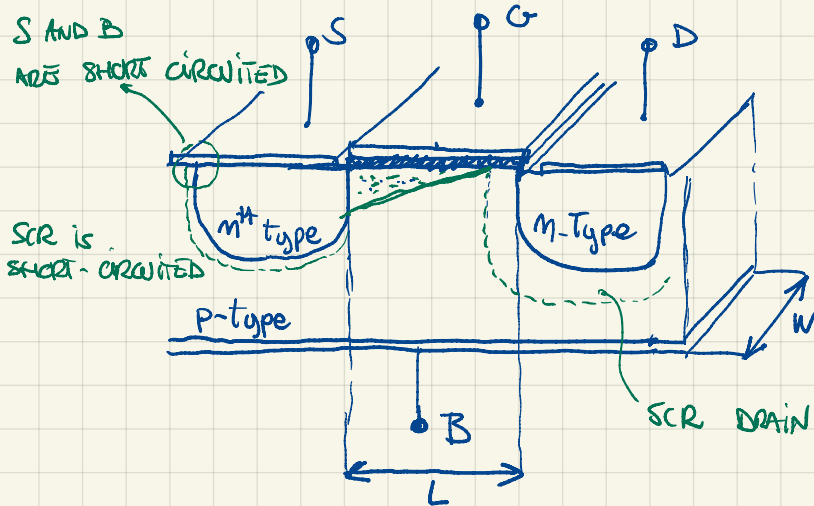
$$C_{CB} \triangleq C_{\mu} = \frac{C_{CB\emptyset}}{\sqrt{4 + \frac{V_{OB}}{\phi_{0CB}}}} \quad \leftarrow \text{CAPACITANCE WHEN } V_{OB} = \emptyset \quad \phi_{0CB} \approx 0.75 \text{ V}$$

AS A RULE OF THUMB $C_{\mu} \ll C_{T0}$

IN CONCLUSION: THE HIGH FREQUENCY MODEL OF THE BJT (SMALL SIGNAL MODEL) IS THE FOLLOWING:



◇ MOSFET

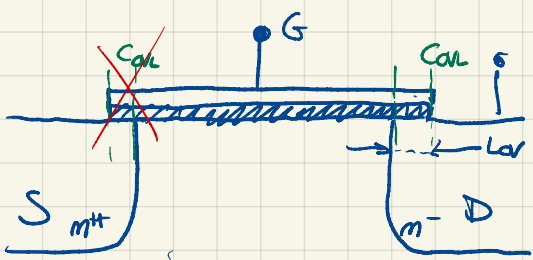


IN THE LINEAR OPERATING REGION (SATURATION) THE DEVICE IS STORING CHARGE AT 3 DIFFERENT LOCATIONS

#1 GATE CHARGE STORAGE $f(V_{GS})$

$$C_{GS} = \frac{2}{3} C_{OX} W L \quad \text{WHERE } C_{OX} = \frac{\epsilon_{SiO_2}}{t_{OX}} \quad [F/cm^2]$$

#2 OVERLAP CHARGE STORAGE



C_{ovL} AT THE SOURCE IS PLACED IN PARALLEL TO C_{GS} WHICH IS TYPICALLY MUCH LARGER \Rightarrow WE NORMALLY NEGLECT THIS COMPONENT

$$C_{GD} = C_{ovL} \cdot W \cdot L_{ov} \quad L_{ov} \in [0.05 \div 0.1L]$$

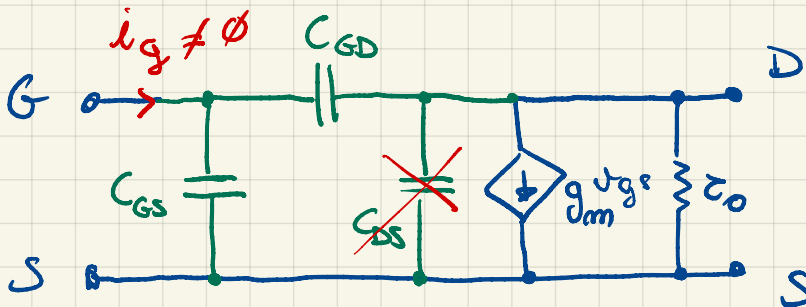
↑
[F/cm²]

#3 D-S JUNCTION CHARGE STORAGE

$$C_{DS} = \frac{C_{DSD}}{\sqrt{1 + \frac{V_{DS}}{\phi_{DSD}}}}$$

AS A RULE OF THUMB $C_{GS} \gg C_{GD} > C_{DS}$

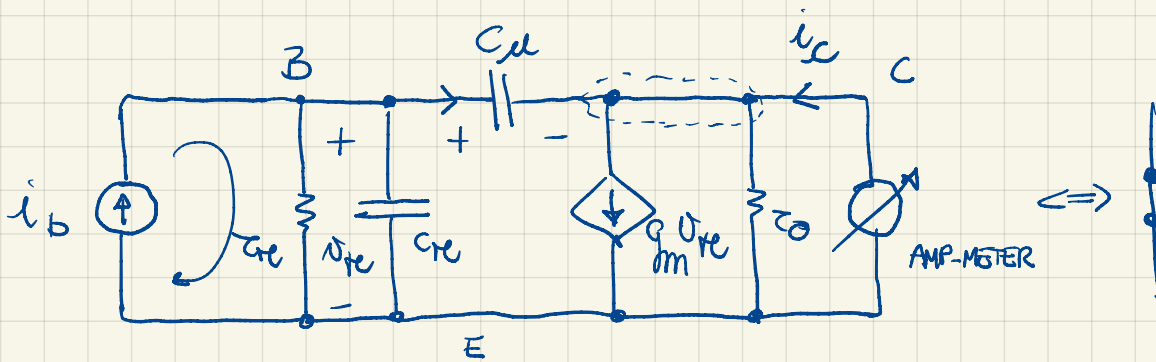
WE CAN NOW DRAW THE SMALL SIGNAL HF MODEL OF THE MOSTFET



THE MODEL HAS REDUNDANT CAPS (THEY FORM A LOOP!), WE SIMPLIFY IT BY NEGLECTING C_{DS} . (THE SMALLEST ONE).

EXAMPLE: LET'S CALCULATE THE SHORT CIRCUIT CURRENT GAIN OF A BJT AS A FUNCTION OF FREQUENCY

$$\beta_o(s) = \frac{I_c(s)}{I_b(s)} \Big|_{V_{CE} = 0}$$



WE ARE CONSIDERING A SPECIFIC OP $\rightarrow r_{re}, C_{re}, C_{\mu}, g_m, r_o$ ARE UNKNOWN

WE NEED TO FIND $i_c = f(i_b) \Rightarrow \beta_o(s)$

(1) KCL AT COLLECTOR NODE

$$i_{C\mu} + i_c = g_m v_{re}$$

(2) CAPACITOR C_{μ} EQUATION

$$i_{C\mu} = s C_{\mu} (v_{re} - \phi) = s C_{\mu} v_{re}$$

(3) KVL AT BASE

$$v_{re} = i_b \cdot \frac{r_{re}}{1 + s r_{re} (C_{re} + C_{\mu})}$$

USING (3) AND (2) INTO (1) WE FIND

DC VALUE OF $\beta_o(s)$
 $\beta_0 = g_m r_{re}$

$$i_b s C_{\mu} \cdot \frac{r_{re}}{1 + s r_{re} (C_{re} + C_{\mu})} + i_c - g_m i_b \frac{r_{re}}{1 + s r_{re} (C_{re} + C_{\mu})} = \phi$$

$$- i_b \frac{-s r_{re} C_{\mu} + \beta_0}{1 + s r_{re} (C_{re} + C_{\mu})} + i_c = \phi$$

RHP ZERO!

$$\beta_o(s) \triangleq \frac{I_c(s)}{I_B(s)} = \beta_0 \frac{\left(1 - s \frac{C_{\mu}}{g_m}\right)}{1 + s r_{re} (C_{re} + C_{\mu})}$$

$$\omega_{ZP} = \frac{g_m}{C_{\mu}}$$

$$\omega_{PB} = \frac{1}{r_{re} (C_{re} + C_{\mu})}$$

$$\frac{\omega_{ZP}}{\omega_{PB}} = \frac{g_m}{C_{\mu}} \cdot r_{re} (C_{re} + C_{\mu}) = \beta_0 \left(1 + \frac{C_{re}}{C_{\mu}}\right) \alpha 10^3$$

$\in [10^2 \div 10^3]$